

Shall we take 13 Mar 2012 off? Discrete FT (FFT) 18.1

Discrete Fourier Transform (FFT)

Fourier modes e^{ikx}

in which k and x run over a finite number N of discrete values.

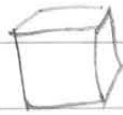
The Fast Fourier Transform (FFT)

evaluates the discrete FT exactly

(to machine precision), and is one of the most widely used numerical techniques.

Q: Where to use FFT?

A: • In hydro simulations of any kind where the background is uniform, and it makes sense to apply periodic boundary conditions:



Periodic box
of N^3 cells.

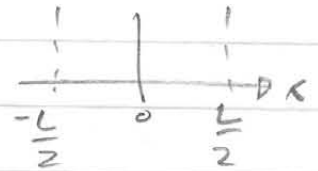
- cosmo sims
- hydro sims
- Imaging processing
 - destriping
 - filtering
- Signal processing
- Audio processing

Discrete FT

Suppose that $a(x)$ is a continuous, real or complex, periodic function.

Wlog, choose periodic interval to be

$$\left[-\frac{L}{2}, \frac{L}{2}\right]$$



Since $a(x)$ is periodic, its Fourier expansion

Fourier modes

$$a(x) = \sum_{m=-\infty}^{\infty} \hat{a}_m e^{2\pi i m x / L}$$

contains only modes $e^{2\pi i m x / L}$ of only integral m .

Q: Wavelength λ of Fourier mode $e^{2\pi i m x / L}$?

$$A: 2\pi i m \lambda / L = 2\pi i \Rightarrow \lambda = \frac{L}{m}$$

Q: Frequency ν ?

$$A: 2\pi i m x / L = 2\pi i \nu x \Rightarrow \nu = \frac{m}{L} = \frac{1}{\lambda}$$

Q: Angular frequency k ?

$$A: 2\pi i m x / L = i k x \Rightarrow k = \frac{2\pi m}{L}$$

Suppose furthermore that $a(x)$ is "smooth" in the specific sense that it contains only modes $|m| \leq \left[\frac{N}{2}\right]$

$$\text{ie } m = 0, \pm 1, \pm 2, \dots, \pm \left[\frac{N}{2}\right]$$

denotes largest integer $\leq N/2$

with the condition that if N is even, then $a_{-\left[\frac{N}{2}\right]} = a_{\left[\frac{N}{2}\right]}$ outermost Fourier coeffs are equal.

$$a(x) = \sum'_{m=-[N/2]}^{[N/2]} \tilde{a}_m e^{2\pi i m x / L}$$

where \sum' means usual sum, except that if N is even, give $\frac{1}{2}$ weight to outermost pair
 ie $\frac{1}{2} (a_{-[N/2]} e^{-2\pi i [N/2] x / L} + a_{[N/2]} e^{2\pi i [N/2] x / L})$

Q: Same as counting one of only $a_{-[N/2]}$ or $a_{[N/2]}$

A: No, as e^i is different

which since $a_{-[N/2]} = a_{[N/2]}$ is same as $\frac{1}{2} a_{[N/2]} (e^{-2\pi i [N/2] x / L} + e^{2\pi i [N/2] x / L})$

Fourier coeffs of these equal.

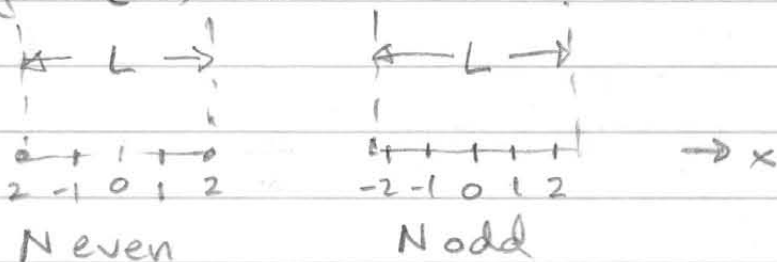
$N = 4$ (even) : $m = 0, \pm 1, \pm 2$

$N = 5$ (odd) : $m = 0, \pm 1, \pm 2$

Let

$a_n \equiv a(x_n)$, $n = 0, \pm 1, \dots, \pm [N/2]$
 denote values of $a(x)$ at

$$x_n = \frac{L}{N} n$$



The sampling theorem asserts that Fourier coeffs \tilde{a}_m are related to a_n by

$$\tilde{a}_m = \frac{1}{N} \sum'_{n=-[N/2]}^{[N/2]} a_n e^{-2\pi i m n / N}$$

$$a_n = \sum'_{m=-[N/2]}^{[N/2]} \tilde{a}_m e^{2\pi i m n / N}$$

constitute a discrete FT pair.

Proof: D14

By definition

$$a_n = \sum'_m \tilde{a}_m e^{2\pi i m n / N}$$

Take

$$\sum'_n a_n e^{-2\pi i m n / N}$$

$$= \sum'_n \left(\sum'_{m'} \tilde{a}_{m'} e^{2\pi i m' n / N} \right) e^{-2\pi i m n / N}$$

$$= \sum'_{m'} \tilde{a}_{m'} \underbrace{\sum'_n e^{2\pi i (m' - m) n / N}}_{\text{can show this} = N \delta_{m'm}}$$

$$= N \tilde{a}_m \quad \checkmark$$

The discrete FT effectively assumes

(a) x is periodic ($\Rightarrow k$ discrete)

(b) k is periodic ($\Rightarrow x$ discrete).

In literature "FT" is often treated as synonymous with "FFT", but it ain't so.

Ex PS8/. Have time series, eg the brightness of a variable AGN.

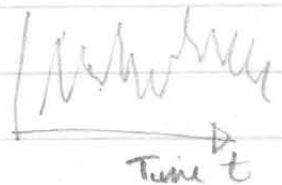
You think the series contains periodicities, so you want to FT.

You only observe for a finite time!

So you decide to FFT. In doing so, you

(a) truncate time series

(b) make it periodic.



Truncation \Rightarrow ringing

Trunc u in real space \Rightarrow ringing in Fourier space.
 u in Fourier space \Rightarrow _____ real space.

Maybe easier to understand latter.

Trunc in Fourier space

\Rightarrow remove freqs above some cutoff

\Rightarrow typically introduces visible oscillations near cutoff freq.

fourier.nb take $N_{seq} = 2^A, 2^B$

Make periodic \Rightarrow aliasing

Make periodic in real space \Rightarrow aliasing in real space
 _____ Fourier \Rightarrow _____ Fourier u .

E.g. in FFT, freq is periodic, so
 frequency $m + iN \equiv$ frequency m
 \uparrow
 any integer

If signal contains freqs higher than your FFT admits, then those freqs will be aliased to lower freq.

Nyquist frequency

The highest frequency that an FFT of N points admits is:

$m = \lfloor \frac{N}{2} \rfloor \leftarrow$ "Nyquist frequency"

or in real units

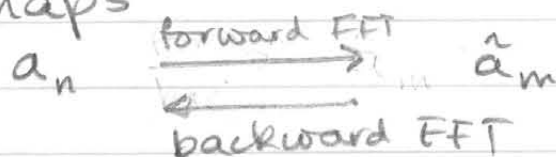
$v = \frac{m}{L} = \frac{\lfloor N/2 \rfloor}{L} \leftarrow$ box size.

Q: Why factor of $1/2$?

A: Essentially because N points encode Re & Im parts of $\lfloor \frac{N}{2} \rfloor$ complex numbers.

Organization of FFT in codes

FFT maps



Typically a_n is a real sequence.

If so, then

$$\tilde{a}_{-m}^* = \tilde{a}_m^*$$

ie \tilde{a}_m is complex, with \tilde{a}_m and \tilde{a}_{-m} effectively encoding the Re and Im parts of \tilde{a}_m :

$$\text{Re}(\tilde{a}_m) = \frac{1}{2}(\tilde{a}_m + \tilde{a}_{-m})$$

$$\text{Im}(\tilde{a}_m) = \frac{1}{2i}(\tilde{a}_m - \tilde{a}_{-m})$$

Code typically inputs real periodic sequence as

$$\begin{array}{ccccccc}
 a_0, a_1, a_2, \dots, a_{\lfloor \frac{N}{2} \rfloor}, & a_{-\lfloor \frac{N}{2} \rfloor + 1}, & \dots, & a_{-1} \\
 & = a_{\lfloor \frac{N}{2} \rfloor + 1} & & = a_{N-1}
 \end{array}$$

and outputs

$$N \tilde{a}_0 = \sum a_n$$

$$N \text{Re} \tilde{a}_1 = \text{Re} \sum a_n e^{-2\pi i n / N}$$

$$N \text{Im} \tilde{a}_1 = \text{Im} \quad "$$

$$N \text{Re} \tilde{a}_{\lfloor \frac{N}{2} \rfloor}$$

$$\begin{array}{l}
 \cos(2\pi n / N) \\
 \sin(2\pi n / N)
 \end{array}$$

Mathematica outputs FFT as complex seq
 $\hat{a}_0, \hat{a}_1, \dots, \hat{a}_{\lfloor \frac{N}{2} \rfloor}, \hat{a}_{-\lfloor \frac{N}{2} \rfloor + 1}, \dots, \hat{a}_{-1}$
 $= a_{\lfloor \frac{N}{2} \rfloor + 1} \quad = \hat{a}_{N-1}$

Mathematica defines FFT

1. with factor $\frac{1}{\sqrt{N}}$ in normalization
2. as complex conjugate of our convention
fourier, nb

With any FFT, check:

1. Normalization.
2. Phase convention.
3. Ordering of sequences.

Remedy for ringing caused by ^{0?} truncation
 Multiply by a window that goes to zero
 at edges of data.

Remedy for aliasing caused by ^{0?} making periodic
 Pad data with the mean of the data.
 Or, pad mean-subtracted or windowed
 data with zeros. Padding to at least
 double length eliminates aliasing between
 start and end of data

Windowing

Convolution theorem (Proof later):

Multiplication in real space \equiv convolution in Fourier space
 ————— Fourier — \equiv ————— real —

agn.nb

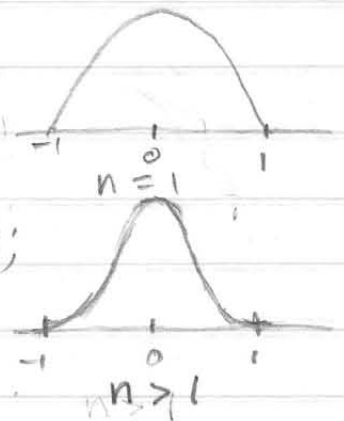
Andrew's favorite windows:

Tou window over interval $[-1, 1]$,

fourier.nb $w(x) = (1 - x^2)^n$

Why?

- non-zero over finite interval;
 - adjustable parameter n gives useful flexibility;
 - As n increases, window goes to zero at edges ever more smoothly.
 - As $n \rightarrow$ large, window \rightarrow Gaussian.
- Note that FT of a Gaussian is a Gaussian.



Q: To get narrow ~~conv~~ window in F space? A: Broad ^{mult} window in real space

Data length

FFT algorithm is fastest when N is a product of small primes (2, 3, 5, ...).

Fastest is $N = 2^n$.

Slowest is $N =$ prime number, in which case FFT reverts to SFT.

Correlation function

Q: Examples? Let $a(x)$ be a real random field that is statistically translation invariant. Its lowest order moments are:

1. Mean $\bar{a} = \langle a(x) \rangle$

2. Covariance, or correlation function

$$\xi(|x'-x|) = \langle \Delta a(x') \Delta a(x) \rangle$$

(function only of: $|x'-x|, \bar{a}^2$)

T_{var} - separation $x'-x$ by translation invariance,

\Rightarrow - abs value $|x'-x|$ by pair exchange symmetry
and pair exchange symmetry

where $\Delta a(x) \equiv a(x) - \bar{a}$, \bar{a} is the mean.

Note: correlation function ξ is often defined to be $1/\bar{a}^2$ of this, so that ξ becomes dimensionless.

power

Power spectrum P_k

Covariance of Fourier modes a_k is

$$\begin{aligned}
 & \langle \Delta a_k^* \Delta a_k \rangle \quad \text{take Fourier series, for simplicity} \\
 & = \left\langle \left(\int_{-\pi}^{\pi} \Delta a(x') \frac{e^{-ik'x'}}{\sqrt{2\pi}} dx' \right)^* \left(\int_{-\pi}^{\pi} \Delta a(x) \frac{e^{-ikx}}{\sqrt{2\pi}} dx \right) \right\rangle \\
 & = \int_{-\pi}^{\pi} \int_{-\pi}^{\pi} \langle \Delta a(x') \Delta a(x) \rangle \frac{e^{ik'x'}}{\sqrt{2\pi}} \frac{e^{-ikx}}{\sqrt{2\pi}} dx' dx \\
 & = \int_{-\pi}^{\pi} \xi(|x'-x|) e^{ik'(x'-x)} d(x'-x) \int_{-\pi}^{\pi} \frac{e^{i(k'-k)x}}{2\pi} dx \\
 & = \sqrt{2\pi} P_{k'} \quad = \delta_{k'k}
 \end{aligned}$$

$\sqrt{2\pi}$ power spectrum
 \uparrow
 $=$ FT of corr fn

$$= \sqrt{2\pi} \delta_{k'k} P_k$$

Q: P_k real, complex, ...?

A: $P_k \propto \langle \Delta a_k^* \Delta a_k \rangle = \langle |\Delta a_k|^2 \rangle$ is +ve real.

Can think of corr fn as matrix

$$\xi(|x'-x|) \equiv \xi_{x'x} \equiv \langle \Delta a_{x'} \Delta a_x \rangle$$

and power spectrum as diagonal matrix

$$P_{k'k} \equiv P_k \delta_{k'k} \equiv \langle \Delta a_{k'}^* \Delta a_k \rangle$$

Q: Why is covariance of Fourier modes diagonal, whereas covariance of real space modes is not?

A: Statistical translation invariance.

Fact that Fourier modes are uncorrelated means power spectrum is greatly superior to corr fn as a statistical tool.

This is true even though ...

The power spectrum P_k is the FT of the corr fu $\xi(x)$, and therefore P_k and $\xi(x)$ contain identical information.

Warning

Conventions for normalization of P_k , disposition of 2π 's, vary.

Discover the convention in your field, and stick to it.

The mean is encoded by zeroth Fourier mode

The zeroth Fourier eigenmode is

$$\frac{e^{ikx^0}}{\sqrt{2\pi}} = \frac{1}{\sqrt{2\pi}} \text{ is constant.}$$

All other Fourier modes are orthog to this.

Thus $\Delta a_k \equiv \text{FT of } \Delta a(x)$

$$\langle k \rangle = \begin{cases} 0 & k=0 \\ a_k & k \neq 0 \end{cases}$$

$$\text{So } \langle \Delta a_k^* \Delta a_k \rangle = \begin{cases} 0 & k=0 \\ \langle a_k^* a_k \rangle & k \neq 0 \end{cases}$$

Estimate power spectrum from FFT of data

Estimate

$$P_k = a_k^* a_k = |a_k|^2$$

for $k = 1, 2, \dots, \left[\frac{N}{2} \right]$.

Shot noise

= self-pair contribution to any statistic.

This definition of shot noise is all you ever need to remember about it.

Shot noise arises whenever a random field is realized by discrete objects

- photons
- galaxies

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Example / PSF.

Estimate power at wavenumber m :

$$\begin{aligned} a_m^* a_m &= \frac{1}{N^2} \left(\sum_{n'} a_{n'} e^{-2\pi i m n' / N} \right)^* \left(\sum_n a_n e^{-2\pi i m n / N} \right) \\ &= \frac{1}{N^2} \sum_{n'} \sum_n a_{n'}^* a_n e^{2\pi i m (n' - n) / N} \end{aligned}$$

contains self-pair contributions from individual x-ray photons.

Necessarily $n' = n$ for self-pairs, but each bin contains many photons.

If $a_n \equiv \#$ phots in bin n , then

$$(a_m^* a_m)_{\text{shot}} = \frac{1}{N^2} \sum_n a_n^2 = \text{constant} \times \text{width of } m$$

If $a_n = W_n$ times ($\#$ phots in bin n) then then

$$(a_m^* a_m)_{\text{shot}} = \frac{1}{N^2} \sum_n W_n^2 a_n$$

Why? Go back to definition of shot noise, and think hard.

Cross-correlation ~~function~~ FT of data
between 2 random fields $a(x)$, $b(x)$

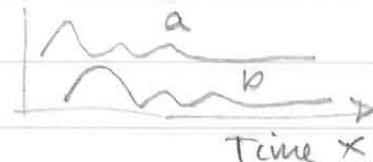
is covariance = $\langle \Delta a(x') \Delta b(x) \rangle$
 $\xi_{\text{cross}}^{ab}(x'-x) = \langle \Delta a(x') \Delta b(x) \rangle$

function only of
 $x'-x$ by translation invariance

Can be that $\xi_{\text{cross}}^{ab}(-x'-x) \neq \xi_{\text{cross}}^{ab}(x'-x)$.

Q: Example where way "no" \neq 0/a?

A: a, b two light curves, where
 1 lags the other.



Q: Example where ξ_{cross}^{ab} is
 a function of $|x'-x|$ only?

A: isotropic 2D field (CMB)
 or 3D (or 3D) "u" (galaxies)

Cross-power

$$P_k^{ab, \text{cross}} = \text{FT of } \xi^{ab}(x)$$

Q: P_k^{ab} real, complex, ...?

A: Complex in general,
 real if $\xi^{ab}(x)$ is func of $|x|$ only.
 May be -ve.

Can look for evidence for lag

Note $\xi^{ab}(-x) \neq \xi^{ab}(x)$
 by looking at $\text{Im}(P_k^{ab})$.

Shot noise in cross-power

If two fields $a(x)$, $b(x)$ share no members (no common photons, galaxies, ...) then the shot noise is ... (0?) zero.

This is how WMAP CMB power spectrum is computed - from cross-powers between maps at distinct wavelength bands - so eliminating uncertainty arising from shot noise subtraction.

If two fields share members, then there will be shot noise from the shared members.

FFT algorithm (sketch) See Numerical Recipes.

Recursively split FFT of length N into 2 FFTs of length $\frac{N}{2}$, if N even.

↑
lowest prime factor,
more generally

$$\hat{a}_m = \frac{1}{N} \sum' a_n e^{-2\pi i m n / N}$$

split sum into even

odd

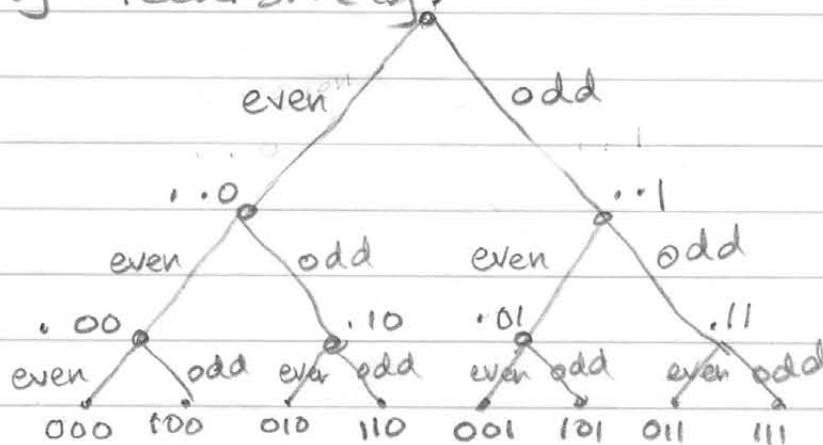
$$= \frac{1}{N} \left[\sum' a_{2n} e^{-2\pi i m 2n / N} + \sum' a_{2n+1} e^{-2\pi i m (2n+1) / N} \right]$$

$= e^{-2\pi i m n / (N/2)}$ $e^{-2\pi i m n / N}$ $e^{-2\pi i m n / (N/2)}$

$$= \frac{1}{N} \left[\sum' a_{2n} e^{-2\pi i m n / (N/2)} + e^{-2\pi i m / N} \sum' a_{2n+1} e^{-2\pi i m n / (N/2)} \right]$$

FFT of length $N/2$ factor FFT of length $N/2$

Apply recursively:



Note bottom layer elements are in bit-reversed order.

Take a_n , write $n = \dots$ in binary, 0s & 1s

Bottom layer is $N=2$ transform:

$$\hat{a}_m = \frac{1}{2} \left[a_0 + a_1 \frac{1}{2} \left(e^{-2\pi i m / 2} + e^{2\pi i m / 2} \right) \right]$$

$$= \begin{cases} 1 & m = 0 \\ -1 & m = \pm 1 \end{cases}$$